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P28330EP

### Claims

5 1. A power control circuitry for controlling the output power level ( $P_{out}$ ) of a signal ( $x(t)$ ) to be transmitted at the output port of a variable-gain power amplifier (105), said power control circuitry (101M+N) comprising  
a current sense loop (101M) with an integrated comparator stage (112'') having a first input port supplied with a reference signal ( $V_{ref}$ ) representing the nominal power level ( $P_{ref}$ )  
10 for the output power ( $P_{out}$ ) and a second input port supplied with a signal from a current sensor (204) which is placed in the power supply line of a variable-gain power amplifier (105), wherein the output signal of said comparator stage (112'') is fed to the power control input port of the variable-gain power amplifier (105),  
characterized by

- 15 – power sensing means (108) for detecting the power of a feedback signal ( $V_{PD}$ ) representing the reflected wave of the signal ( $x(t)$ ) to be transmitted, and  
– a feedback loop (101N) for feeding said reference signal ( $V_{ref}$ ) derived from said feedback signal ( $V_{PD}$ ) and a reference ramp signal ( $V_{ramp}$ ) to the first input port of the comparator stage (112'') in order to increase the radiated power ( $P_{out}$ ) of said signal ( $x(t)$ ) in  
20 case a transmit antenna (110) is mismatched to the variable-gain power amplifier (105).

2. A power control circuitry according to claim 1,  
characterized by

signal processing means comprising

- 25 – a multiplier (301b) for multiplying a processed version ( $K \cdot G_{OP} \cdot V_{PD}$ ) of the feedback signal ( $V_{PD}$ ) by the reference ramp signal ( $V_{ramp}$ ),  
– a summation element (301a), used for adding the output signal ( $V_{ramp} \cdot K \cdot G_{OP} \cdot V_{PD}$ ) of the multiplier (301b) to the reference ramp signal ( $V_{ramp}$ ), thereby yielding said reference signal ( $V_{ref}$ ).

3. A power control circuitry according to claim 1,

characterized by

digital signal processing means (201C) comprising a multiplication element (301b') for

multiplying a gain factor ( $\chi := 1 + K \cdot G_{OP} \cdot V_{PD}$ ) supplied by a gain factor control unit (301c)

5 by the reference ramp signal ( $V_{ramp}$ ), wherein  $K$  is a normalization factor (in  $V^{-1}$ ) and  $G_{OP}$  denotes the gain factor of an operational amplifier (303) in said feedback loop (101N), thereby yielding said reference signal ( $V_{ref}$ ).

4. A power control circuitry according to anyone of the claims 1 to 3,

10 characterized by

decoupling means (106) at the output port of the variable-gain power amplifier (105) for providing a feedback signal ( $V_{PD}$ ).

5. A power control circuitry according to claim 4,

15 characterized in that

said decoupling means (106) is realized as a directional coupler (106') or a circulator (106'').

6. A method for stabilizing the power level ( $P_{out}$ ) of a signal ( $x(t)$ ) to be transmitted at the  
20 output port of a variable-gain power amplifier (105),

said method being characterized by the following steps:

- detecting (S1) the voltage level ( $V_{PD}$ ) of a feedback signal which represents the reflected wave of said signal ( $x(t)$ ),
- calculating (S1A) a reference signal ( $V_{ref}$ ) representing the nominal power level ( $P_{ref}$ ) for  
25 the output power ( $P_{out}$ ) of the RF output signal ( $x(t)$ ) as a function of a reference ramp signal ( $V_{ramp}$ ) and said feedback signal ( $V_{PD}$ ),
- feeding (S2) the obtained reference signal ( $V_{ref}$ ) to a first input port of a comparator stage (112'') in the feedback chain of the current sense loop (101M),
- feeding (S4) a signal representing the DC supply current ( $I_{PA}$ ) of the variable-gain power  
30 amplifier (105) to a second input port of said comparator stage (112''),
- comparing (S5) the voltage level of the signal derived from said voltage drop ( $U_{RM}$ ) with the voltage level of said reference signal ( $V_{ref}$ ),

- feeding (S6) a signal being a function of the difference between the signal derived from said voltage drop ( $U_{RM}$ ) and the calculated reference signal ( $V_{ref}$ ) to a first input port of the power amplifier (105), and
- adjusting (S7) the current power level ( $P_{out}$ ) by amplifying the difference between the output signal of said comparator stage (112'') and the signal ( $x(t)$ ) to be transmitted before being amplified at a second input port of the variable-gain power amplifier (105).

7. A method according to claim 6,  
characterized in that

- 10 the step (S1A) of calculating said reference signal ( $V_{ref}$ ) comprises the following steps:
- multiplying (S1a') a processed version ( $K \cdot G_{OP} \cdot V_{PD}$ ) of the feedback signal ( $V_{PD}$ ) by the reference ramp signal ( $V_{ramp}$ ) and
  - adding (S1a'') the output signal ( $V_{ramp} \cdot K \cdot G_{OP} \cdot V_{PD}$ ) of the multiplication step (S1a') to the reference ramp signal ( $V_{ramp}$ ), thereby yielding said reference signal ( $V_{ref}$ ).

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8. A method according to claim 6,  
characterized in that

- the step (S1A) of calculating said reference signal ( $V_{ref}$ ) comprises the step of multiplying (S1b) a gain factor ( $\chi := 1 + K \cdot G_{OP} \cdot V_{PD}$ ), which is supplied by a gain factor control unit (301c), by the reference ramp signal ( $V_{ramp}$ ), thereby yielding said reference signal ( $V_{ref}$ ).

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9. A wireless telecommunication device,  
characterized by

- 25 a mobile RF transmitter (300a, 300b or 300c) comprising a power control circuitry (101M+N) according to anyone of the claims 1 to 5.